# First results of T2K-nd280 Front End Electronics performance with GM-APDs

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- T2K and the 280m near detectors
- overview of the Trip-t Front End Board (TFB)
- ➡ Measurements with MPPC 100/400 pixels
  - Charge spectra
  - Timestamping
  - Voltage trim functionality tests

## Tokai to Kamioka

### 12 countries, 62 institutions, ~ 350 people



- ⇒ 2009 Phase I :  $\theta_{13}$ ,  $\theta_{23}$ ,  $\Delta m^2_{23}$ 
  - J-PARC : 0.75 MW 30 GeV
  - SK-III : 22.5 kT FV, full PMT coverage
- → 2015 Phase II :  $\theta_{13}$ ,  $\delta_{CP}$ ?
  - J-PARC : 4MW 50 GeV
  - HyperK : 1 MT scale



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### 280m near detectors complex



### Scintillators readout constraints

### Magnetic field

• UA1 magnet will be operated at B = 0.2 T

### Low light yield

- In scintillators sub-systems ~ 10-15 p.e./MIP/cm expected
- Very tight space constraints
  - small space for readout
- High number of channels
  - ~60000 total
- Detector in operation for 5 years
  - Low maintance is desirable

# GMAPD is only candidate that met (almost) all requirements

### Scintillators Readout

Scintillator bar readout cut view (ECAL)





- Scintillator + Wavelength shifting fibre + GMAPD
  - Kuraray Y-11 1 mm diameter WLS fibre
- Tight readout space in UA1 magnet
- GMAPDs have individual connector
- Good coupling crucial to minimise light loss at fibre end

### ND280 Electronics overview



## TRIP-t Front end board (front view)

#### miniature coax connectors for photo-sensors

#### 12 layers – 6 routing, 6 power/GND



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regulators

**10 bit ADCs** 

### TRIP-t Front end board (back)



### 8 x 8 channel HV trim DACs - 5V trim range on every channel

### TRIP-t parameters

- ➡ 32 synchronous channels
- Adjustable Integration window
  - 50 nsec to many musec
  - reset time can also be adjusted, 50 nsec minimum
- ➡ Adjustable gain
  - saturation at 3000 fC
  - noise < 1fC
- Buffer depth 23 timeslices
- Timestamp discriminator threshold for each TRIP-t
  - 1 timestamp per channel/integration window
- Timestamp generation from 400MHz TDC (FPGA)
  - 2.5 ns resolution

## J-PARC Spill structure

#### **Spill Structure**



- ➡ 8(15) bunches per spill
- → 4(2) muon lifetime after spill active period (90(80)%) active)
  - translates to 50-70% in Michel electron tagging efficiency

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## TRIP-t front end



- Only pre-amp affects signal feeding discriminator
  - no fine control (x1 or x4)
- discriminator threshold Vth
  - common to all channels on chip
- Analogue bias settings
  - gain, Vth, etc
  - programmable via serial interface

### **GMAPD**-TFB connection



- HVglobal : common to all GMAPD channels on the TFB
- HV Trim : 5V individual bias voltage adjustment
- HV Trim applied to coax sheath AC coupled to GRD

### Interconnections



- ➡ Miniature coaxial cable (HRS)
  - stands voltage input up to 100V
- min coax connectors on top surface
- Gain splitting and bias components on bottom surface
- Electric fan-in to TRIP-t inputs on internal layer to avoid pick-up





### **TFB** Pedestal and noise



average pedestal value for all 4 chips

small systematic chip-to-chip differences

1 p.e. ~ 10 ADC units (for 5x105 photo-sensor gain)

noise ~ 1 ADC unit small difference in noise between high and low gain channels

## TFB linearity



- 64 high (red) and low (blue) channels (4 TRIP-t's, 16 hi/lo channels per chip
- Behavior ~ identical to single TRIP-t chip
- ➡ 5½ channel spread attributed to gain setting component tolerances
- Calibration required to correct nonlinearities



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### TFB Discriminator settings



discriminator turn-on curves for all 16 channels from 1 trip-t

measurement procedure:

inject fixed external charge (200 triggers) sweep discriminator thresh. voltage Vth count no. of times discriminator fires (no. of timestamps)

3 separate measurements for 1.5, 2.5 and 3.5 p.e. equivalent external injected charge (assuming 5x10<sup>5</sup> electrons/p.e.)

channel-to-channel spread better than that previously measured on single chip test board possibly attributable to shielded layout of fan-in tracking between input connectors and trip-t I/Ps?

### Measurements using GMAPD

### Motivation :

- Check behavior with T2K GMAPD candidates
- Identify problems with prototype
- Start developing large scale sensor QA methods using the TFB

### ➡ TFB prototypes received beginning of June

- Currently under tests
- TFB firmware well advanced, almost all functionalities being implemented

### Test setup



➡ MPPC 400 and 100 pixels (S1036211XXX-C)

- Temperature T=25°C
- 400 pixels : G ~ 7x10<sup>5</sup> DCR ~ 400kHz
- 100 pixels : G ~ 1x10<sup>6</sup> DCR ~ 500kHz
- NANOLED source, 1 ns pulse width
- TFB settings
  - 10 integrations window : 250 ns
  - reset period : 100 ns



### MPPC 400pix ADC spectrum



### MPPC 400pix low LED ~ 4-5pe

 $T = 25^{\circ}C, V = 69.81V$ 



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### MPPC 400pix medium LED ~ 50pe



### MPPC 100pix low LED ~ 4-5pe



### MPPC 100pix high LED ~ 15pe



## Effect of large signals

# Large signals induce drop in voltage line :

- Total charge integrated decreases in following timeslice
- gain drops by 10% for 100 p.e. signal
- May affect timestamping
- HV line recovery time is in the order of few µs
  - Gain drop can be corrected
  - HV line resistor values could be adjusted but signal in next timeslice after very large one will be rare



#### Pedestal mean value HiGain channel 10 timeslice

### Time Stamping



- Timestamp value is correlated to total charge integrated
- Discriminator timestamp reliable only when Q >> Qvth
- What performance to expect with real GMAPD pulse integration ?



## Discriminator Time walk with MPPC I

- ➡ HPK-S1036211050-C (400 pixels)
- → Threshold Vth = 1.5 p.e.
- → Acquired 10000 triggers for 4 different LED intensity (0 to ~40 p.e.)
  - Readout ADC data and timestamp
  - Cut on 2nd timeslice
  - Correlation shows sum of all 4 measurements for N<sub>timestamp</sub> > 0
- Time resolution ~ 1 ns for signal > 10 p.e.
- Very promising timing performance with GMAPD response



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## Discriminator Time walk with MPPC II

HPK-S1036211100-С (100 pixels)

- Threshold Vth = 3.5 p.e.
- Same measurement as before but Vth set according to increase in gain
- Good time resolution but larger spread of signal due to long pixel decay time (~100 ns) and possibly afterpulse



- Preliminary results suggests that a time resolution of 1 ns or better is achievable
  - study in progress
- Ultimately time performance will depend on light yield and threshold setting for each sub-system
  - Need full detection chain test with cosmic muons
  - Optimum threshold setting depends on dark count rate and light yield

### ➡ Motivation :

- Sensor parameters measurements (QA) using ADC spectrum
  - gain
  - dark count rate,
  - pixel crosstalk, afterpulse
- T2K run : inter-calibration and correction for gain variations

➡ TFB has 10bit DAC : +5V, 20 mV voltage resolution

## Voltage scan

### ➡ 80 mV steps, pedestal run

- 4x more points possible with DAC
- Gain extracted using peak to peak method
- Dark count rate estimation using ratio : 0.5p.e./Nevent
- Study voltage scans in more details to ensure good accuracy of voltage control



### Summary and schedule

- Early study of the TFB prototype shows very good performances with MPPC devices
  - good signal charge and time resolution
- TFB functionalities under tests
  - Voltage scan using HV trim works
- ➡ TFB final modifications before end 2007
- Production of TFB and Back end board Apr 2008
- GMAPD delivery and tests planned in early 2008 for most nd280 sub-systems
- ➡ INGRID commissioning Jan 2009
- T2K starts Apr 2009
- nd280 commissioning Oct 2009
- nd280 data taking Nov 2009

### Chip to Chip variations





repeat previous measurement for other 3 trip-t's (2.5 p.e. Qin only)

spread ~ same for all 4 chips

small systematic chip-chip offset, but can program Vth individually for each chip anyway

### TFB channel crosstalk

data from one trip-t, injecting external charge on one channel

2.5 pC



for 500 p.e. signal (40 pC) get ~ 1 p.e. signal in neighbouring hign gain channels, and ~1 p.e. depression of pedestals in other high gain channels (~ 0.2 % effects)



### Programmable Integration/Reset



preamp integration/reset time independently programmable



reset period